

Distributed Scalar Quantizers for Subband Allocation

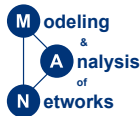
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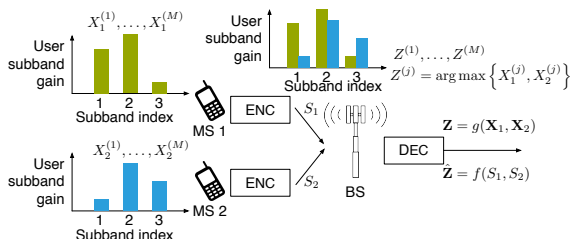
Conference on Information Sciences & Systems
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March 20th, 2014

Outline

- 1 Introduction
- 2 Problem Model
- 3 Optimal Scalar Quantizer Design
 - Homogeneous Scalar Quantizers
 - Heterogeneous Scalar Quantizers
- 4 Results
- 5 Conclusions

Motivation

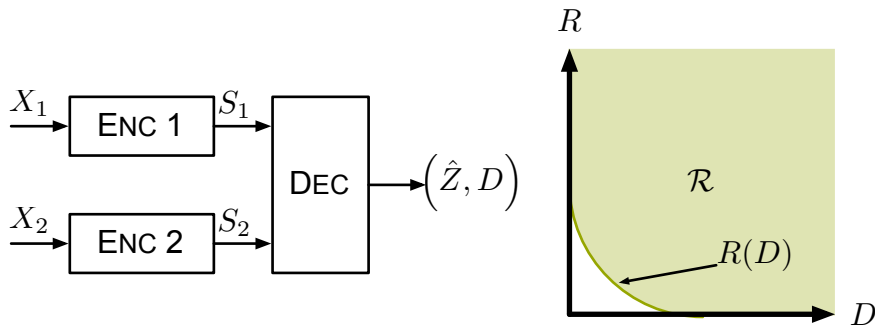
- Subbands in an OFDMA system must be assigned to a unique MS
- AMC: BS wants MS w/ best channel & the gain on that channel
- *Rateless codes* (e.g., ARQ): BS wants MS w/ best channel



- BS does not need to reproduce MS local state
- Trade-off in feedback overhead & system efficiency

Related Work

CEO—Indirect Distributed Lossy Source Coding



- $S_i \in \{1, \dots, 2^{nR_i}\}$
- $Z = g(X_1, X_2)$, $\hat{Z} = f(S_1, S_2)$
- $D = \mathbb{E} [d(Z, \hat{Z})]$, $R = R_1 + R_2$
- \mathcal{R} achievable (R, D) pairs
- $R(D)$ smallest R s.t. $(R, D) \in \mathcal{R}$

Cover & Thomas 2006 ([1]) and El Gamal & Kim 2011 ([2])

Related Work

Distributed Functional SQ & Layered Architectures

Zamir & Berger (1999) [3]

- Lossy, continuous sources
- Lattice quantizers w/ Slepian-Wolf (SW) coding
- Optimal **asymptotically in rate**

Servetto (2005) [4]

- Lossy, **discrete** sources
- **Scalar** quantizer w/ SW coding
- Optimal **for all rates**

“Layered” Achievable Scheme

Scalar Quantizers at each user followed by entropy coding

Wagner et al. (2008) [5]

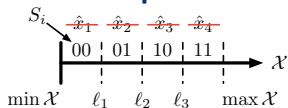
- Lossy (**MSE**), **Gaussian** sources
- Vector quantizers w/ SW coding
- Optimal for all rates

Misra et al. (2011) [6]

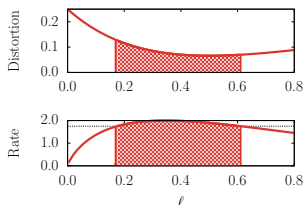
- Lossy (MSE), **function** of sources
- High rate regime
- Optimal asymptotically in rate

Summary of Contributions

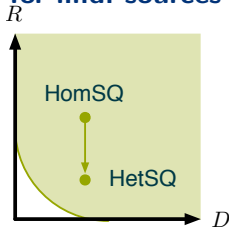
1. Distortion optimal HomSQs



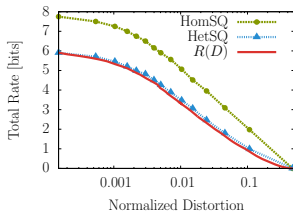
2. Entropy-constrained HomSQs



3. HetSQs superior to HomSQs for i.i.d. sources



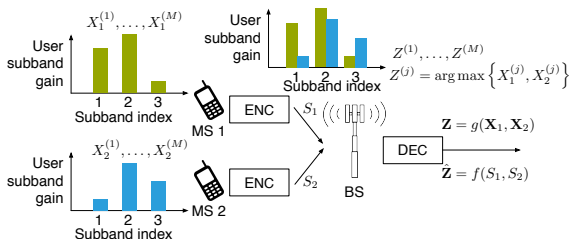
4. HetSQ can be close to fundamental limit



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Problem Model & Notation



- X_i i.i.d. chan. capacity for MS i
- Z optimal subband allocation
- \hat{Z} estimated subband allocation
- S_i coded message from MS i
- R_i rate achieved by MS i
- d distortion

$$Z = \arg \max_i \{X_i : i = 1, 2\}$$

$$d(Z, \hat{Z}) = X_Z - X_{\hat{Z}}$$

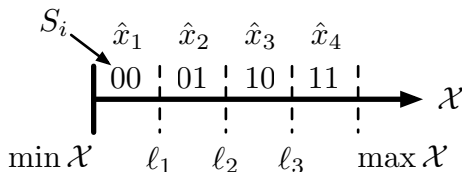
We focus on two users w/ single subband

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Scalar Quantizers

- K -bin SQ parameterized by
 - *decision boundaries* l_i
 - *reconstruction points* \hat{x}_i
- Designed to meet distortion and/or rate constraints
- **Encoding:** report the index of the bin containing X_i
- **Decoding:** map bin index to reconstruction points



$$l_0 \triangleq \min \mathcal{X} \quad l_K \triangleq \max \mathcal{X}$$

X_i MSi's **channel capacity**
 \mathcal{X} **support set** for r.v. X_i
 S_i MSi's **message** to BS

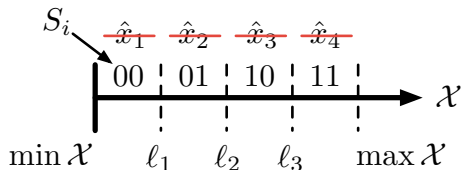
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Minimum Distortion Scalar Quantizers

Obs: Not reproducing local state \Rightarrow reconstruction points not needed

HomSQ: Both users have identical quantizer decision boundaries



- Distortion is a function of ℓ
- Select ℓ to minimize $D(\ell)$

Theorem

If ℓ is an optimal HomSQ then there exists $\mu \geq 0$ such that

$$f_X(\ell_k) \int_{\ell_{k-1}}^{\ell_{k+1}} (\ell_k - x) f_X(x) dx - \mu_k + \mu_{k+1} = 0$$

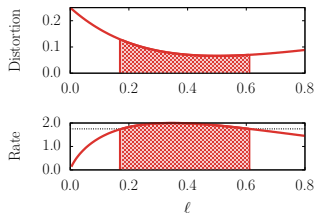
$$\mu_k (\ell_{k-1} - \ell_k) = 0.$$

Entropy Constrained Scalar Quantizers

- Rate is also a function of ℓ
- Select ℓ to minimize $D(\ell)$ w/
an upper-limit R_0 on rate

$$p_k \triangleq \mathbb{P}(S_i = k) = F_X(\ell_k) - F_X(\ell_{k-1})$$

$$R_{HomSQ}(\ell) = H(S_1) + H(S_2) = 2H(s)$$



Problem is non-convex

Theorem

If ℓ is an optimal ECSQ, then there exists $\mu \geq 0$ and $\mu_R \geq 0$ such that

$$f_X(\ell_k) \left(\int_{\ell_{k-1}}^{\ell_{k+1}} (\ell_k - x) f_X(x) dx + 2\mu_R \log_2 \left(\frac{p_{k+1}}{p_k} \right) \right) - \mu_k + \mu_{k+1} = 0$$

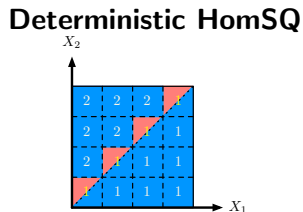
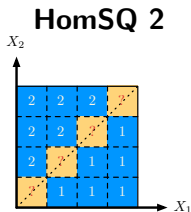
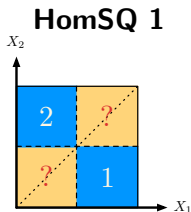
$$\mu_k(\ell_{k-1} - \ell_k) = 0 \text{ and } \mu_R(R_{HomSQ}(\ell) - R_0) = 0.$$

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From HomSQs to HetSQs

Some Insights



- Obvious for $S_1 \neq S_2$
- Flip a coin for $S_1 = S_2$
- Distortion only along diagonal

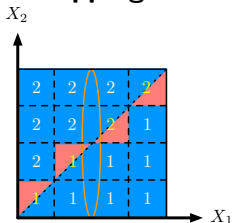
- $\arg \max$ not symmetric
- Distortion *is*
- Same distortion for a fixed mapping along diagonal

From HomSQ to HetSQ

Rate Reduction

Obs: All mappings have the same distortion; some have better total rate

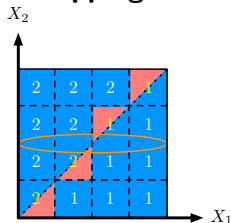
Mapping One



$$R_{HetSQ}^{(1)} \leq R_{HomSQ}^{(1)}$$

Follows from subadditivity of $t \log t$

Mapping Two



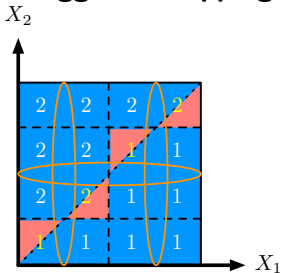
$$R_{HetSQ}^{(2)} \leq R_{HomSQ}^{(2)}$$

Theorem

For an optimal HomSQ ℓ^ that achieves a distortion $D(\ell^*)$, there exists a HetSQ that achieves the same distortion but at a lower rate.*

Staggered HetSQ

Staggered Mapping



$$R_{\text{HetSQ}}^{(1)} \leq R_{\text{HomSQ}}^{(1)}$$

$$R_{\text{HetSQ}}^{(2)} \leq R_{\text{HomSQ}}^{(2)}$$

Theorem

For a HetSQ, if there exists an quantization interval for a user that is completely contained in the quantization interval for another user, then the quantizer is not optimal.

Design of HetSQ

- 1: Select the total # of bins K
- 2: Design optimal HomSQ boundaries ℓ_{HomSQ}
- 3: Assign $\ell_{\text{HomSQ}}^{(k)}$ to MS1 if k odd; else, MS2

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Example 1

Uniform(a, b) Channel Capacity

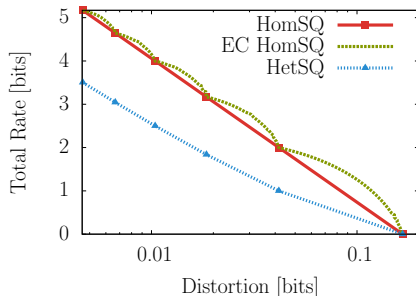
For $k = 1, \dots, K - 1$, the optimal quantizer is given as

$$\ell_k^* = \frac{aK + (b - a)k}{K}, \quad \mu_k^* = 0$$

HetSQ—*No free lunch*; consider

- $K = 3$: 42.1% fewer bits
 - MS2 scheduled w.p. 0.556
 - $R_{HetSQ}^{(2)} = R_{HetSQ}^{(1)}$
- $K = 4$, 37.5% fewer bits
 - MS2 scheduled w.p. 0.500
 - $R_{HetSQ}^{(2)} = 0.67R_{HetSQ}^{(1)}$

Uniform(0, 1) & $K = 1, \dots, 6$



Example 2

Exponential (Exp(λ)) Channel Capacity

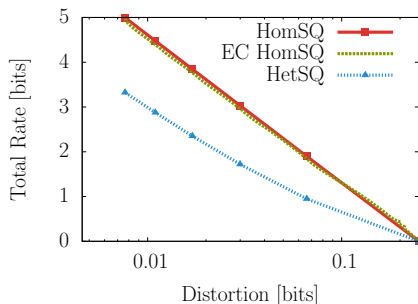
Let $w_k = \lambda l_k$; then the optimal quantizer is given as

$$w_k^* = \frac{-e^{-w_{k-1}^*}(1 + w_{k-1}^*) + e^{-w_{k+1}^*}(1 + w_{k+1}^*)}{(e^{-w_{k+1}^*} - e^{-w_{k-1}^*})}$$

HetSQ—*No free lunch*; consider

- $K = 3$: 43.3% fewer bits
 - MS2 scheduled w.p. 0.560
 - $R_{HetSQ}^{(2)} \approx 0.73R_{HetSQ}^{(1)}$
- $K = 4$: 38.9% fewer bits
 - MS2 scheduled w.p. 0.534
 - $R_{HetSQ}^{(2)} \approx 0.67R_{HetSQ}^{(1)}$

Exp(2) & $K = 1, \dots, 6$



Example 3: LTE CQI

Discrete Uniform Channel Capacity

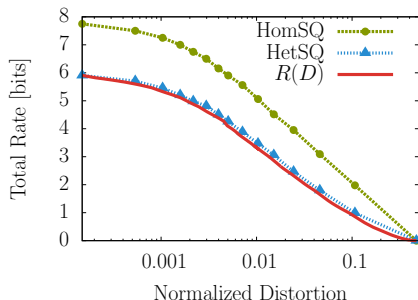
Q: Can we compare HomSQ and HetSQ to a fundamental limit?

A: R-D function computed from Berger-Tung bound

Berger-Tung inner & outer bounds coincide for independent sources

- 16 CQI levels in LTE [7]
- $D = 0$ for $R_1 + R_2 \leq 8$ bits
- HetSQ within 0.124 bits
- HomSQ within 1.622 bits

SQ w/ SW coding is optimal for recovery of discrete sources (Servetto 2005) [4]



Produced w/ help from Gwanmo Ku & Jie Ren

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Review of Contributions

- 1 Distortion optimal HomSQs
- 2 Entropy-constrained distortion optimal HomSQs
- 3 Simple HetSQs achieve same distortion w/ lower rate as best HomSQs
- 4 HetSQ for discrete uniform distribution is close to fundamental limit

Future Work








- Fundamental limit for continuous sources
- Generalize 2-user to N -user
- Consider VQs for multiple subbands
- Investigate low-rate performance as $N \rightarrow \infty$

Acknowledgments



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